

Algorithm for Computation of DCT and its Implementation using a Systolic Architecture

Anamika Jain, Neeta Pandey



Abstract: In this paper a new algorithm for computing N-point DCT, where $N=4r$, $r>1$ is presented. A new algorithm has been derived that can compute the 1D DCT and it is realized in systolic array that utilizes identical processing elements (PE's). The proposed approach can be used to obtain other transform like Discrete Sine Transform (DST), Discrete Hartley Transform (DHT). The suggested algorithm requires reduced number of multiplications as compared to the other methods of computing DCT. This suggests structure meets the architectural challenge and it is simple, regular design and cost-effective for special-purpose system.

Keywords: Processing Element, Systolic architecture, DST, DCT, and DHT.

I. INTRODUCTION

The Discrete Cosine Transform (DCT) play an important role in many Digital signal processing(DSP) applications, since it is good alternative of DFT. It is used in most digital media, including digital images such as in JPEG where some high-frequency components can be discarded as they are redundant, digital video, digital audio (MP3), digital radio (AAC+ and DAB+), digital television (SDTV, HDTV). DCTs are also important to reduce the usage of network bandwidth and also used in finding the numerical solution of partial differential equations using spectral method [1-4]. However, to meet the demand of real time applications, dedicated VLSI implementation of DCT is inevitable [5-7]. Different style of implementation has been reported in the literature including multiplierless based design [8], distributed arithmetic ROM based design [9,10]. Different algorithms have been reported to compute DCT such as FDCT (including Radix 2 and 4) [11-12], Recursive algorithm [13-15]. Among them radix 4 algorithm approach have the features of fast results. There are essentially two ways to build a fast computer system. One is to use concurrency, and the other is to use fast components. Systolic structures are concurrent structures and are useful for implementing a diversity of parallel algorithms [16-21].

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These features are useful for us to develop an efficient design for 1D, N length DCT algorithm as well as systolic architecture. In the presented work, a new algorithm and a concurrent structure for DCT computation is suggested and its systolic architecture is also suggested in the paper. Paper is organized as follows next section (Section II) presents the new algorithm for computing DCT Section III shows the systolic architecture for the presented algorithm. To get more clear explanation computations of DCT N=16 points sequence is presented in section IV. Conclusion and Compression are discussed in section V.

II. PROCEDURE FOR PAPER SUBMISSION

II. A New algorithm for DCT: The Discrete Cosine

Transform (DCT)

$$X[K] = \sum_{n=0}^{N-1} x[n] \cos\left(\frac{\pi k(2n+1)}{2N}\right) \quad k = 0,1,2,\dots,N-1 \quad (1)$$

Dividing eq.(1) into four groups

$$X[K] = \sum_{n=0}^{\frac{N}{4}-1} x[n] \cos\left(\frac{\pi k(2n+1)}{2N}\right) + \sum_{n=\frac{N}{4}}^{\frac{N}{2}-1} x[n] \cos\left(\frac{\pi k(2n+1)}{2N}\right) + \sum_{n=\frac{N}{2}}^{\frac{3N}{4}-1} x[n] \cos\left(\frac{\pi k(2n+1)}{2N}\right) + \sum_{n=\frac{3N}{4}}^{N-1} x[n] \cos\left(\frac{\pi k(2n+1)}{2N}\right), \quad k = 0,1,2,\dots,N-1 \quad (2)$$

$$X[K] = \sum_{n=0}^{\frac{N}{4}-1} \left\{ \begin{aligned} &x[n] \cos(\theta_k) + x\left[n + \frac{N}{4}\right] \cos\left(\theta_k + \frac{\pi K}{4}\right) \\ &+ x\left[n + \frac{N}{2}\right] \cos\left(\theta_k + \frac{\pi K}{2}\right) + \\ &x\left[n + \frac{3N}{4}\right] \cos\left(\theta_k + \frac{3\pi K}{4}\right) \end{aligned} \right\} \quad (3)$$

Let

$$x[n] = A, \quad x\left[n + \frac{N}{4}\right] = B, \quad x\left[n + \frac{N}{2}\right] = C, \quad x\left[n + \frac{3N}{4}\right] = D$$

$$\text{And } \theta_k = \frac{\pi K(2n+1)}{2N}$$



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Equation 3 can be written as

$$X[K] = \sum_{n=0}^{N-1} \left\{ \begin{array}{l} A \cos(\theta_k) + B \cos\left(\theta_k + \frac{\pi K}{4}\right) + \\ C \cos\left(\theta_k + \frac{\pi K}{2}\right) + D \cos\left(\theta_k + \frac{3\pi K}{4}\right) \end{array} \right\} \quad (4)$$

$$X[K] = \sum_{n=0}^{N-1} \left\{ \begin{array}{l} \cos(\theta_k) \left[A + B \cos\left(\frac{\pi K}{4}\right) + C \cos\left(\frac{\pi K}{2}\right) \right] + \\ D \cos\left(\frac{3\pi K}{4}\right) \\ \sin(\theta_k) \left[B \sin\left(\frac{\pi K}{4}\right) + C \sin\left(\frac{\pi K}{2}\right) \right] + \\ D \sin\left(\frac{3\pi K}{4}\right) \end{array} \right\} \quad (5)$$

Let $X[K] = X[4p + q]$

Where $p=0$ to $N/4-1$, and $q=0, 1, 2, 3$

Case 1: $q=0$

$$X[4p] = \sum_{n=0}^{N-1} \left\{ \cos(\theta_{p,q}) \left[A + B \cos(\pi p) + C \cos(2\pi p) + D \cos(3\pi p) \right] \right\} \quad (6)$$

i.e.

$$X[4p] = \sum_{n=0}^{N-1} \left\{ \cos(\theta_{p,q}) \left[A + C + (-1)^p (B + D) \right] \right\} \quad (7)$$

Case2: $q=1$

Equation 5 reduces to

$$X[4p+1] = \sum_{n=0}^{N-1} \left\{ \begin{array}{l} \cos(\theta_{p,q}) \left(A + \frac{(-1)^p}{\sqrt{2}} (B - D) \right) + \\ \cos\left(\theta_{p,q} + \frac{\pi}{2}\right) \left(C + \frac{(-1)^p}{\sqrt{2}} (B + D) \right) \end{array} \right\} \quad (8)$$

Similarly

Case3: $q=2$

$$X[4p+2] = \sum_{n=0}^{N-1} \left\{ \begin{array}{l} \cos(\theta_{p,q}) (A - C) + \\ (-1)^p \cos\left(\theta_{p,q} + \frac{\pi}{2}\right) (B - D) \end{array} \right\} \quad (9)$$

And Case 4: $q=3$

$$X[4p+3] = \sum_{n=0}^{N-1} \left\{ \begin{array}{l} \cos(\theta_{p,q}) \left(A - \frac{(-1)^p}{\sqrt{2}} (B - D) \right) - \\ \cos\left(\theta_{p,q} + \frac{\pi}{2}\right) \left(C - \frac{(-1)^p}{\sqrt{2}} (B + D) \right) \end{array} \right\} \quad (10)$$

Therefore, for even value of $q=0,2$ generalized equation is :

$$X[4p+q] = \sum_{n=0}^{N-1} \left\{ \begin{array}{l} \cos(\theta_{p,q}) \left(A + (-1)^{q/2} C \right) + \\ (-1)^p \cos\left(\theta_{p,q} + \frac{q\pi}{4}\right) \left(B + (-1)^{q/2} D \right) \end{array} \right\} \quad (11)$$

Similarly generalized equation for odd values of q is:

$$X[4p+q] = \sum_{n=0}^{N-1} \left\{ \begin{array}{l} \cos(\theta_{p,q}) \left(A - \frac{(-1)^{3+q/2} (-1)^p}{\sqrt{2}} (B - D) \right) + \\ \cos\left(\theta_{p,q} + \frac{\pi}{2}\right) (-1)^{3+q/2} \left(C - \frac{(-1)^{3+q/2} (-1)^p}{\sqrt{2}} (B + D) \right) \end{array} \right\} \quad (12)$$

Now equations (11) and (12) are the generalized equations for DCT coefficient computation and has been used realization of DCT systolic structure.

III. BLOCK DIAGRAM OF THE PROPOSED ALGORITHM

In designing special-purpose systems, cost-effectiveness and Fast solutions are always main concern. Fast computations can be achieved by using fast algorithms and costs can be reduced by the use of suitable architectures. Great saving can be achieved by decomposing a structure into a few simple substructures, which are used repetitively with simple interfaces. This is especially true for VLSI designs where a single chip comprises hundreds of thousands of components. Systolic architectures are concurrent structures- that can map high-level computations into hardware structures. In a systolic system, data flows in a rhythmic fashion, passing through many processing elements before it returns to memory and can achieve high throughput with balance memory bandwidth. Figure 1 shows a basic block diagram of the proposed algorithm where Pre-processing of the input data is performed to reduce the number of multiplications required to calculate the DCT coefficients.

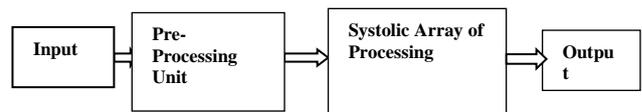
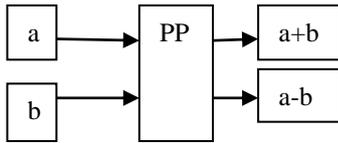


Fig 1. Block diagram of the proposed algorithm

Pre-processing units contains only adders and Subtractors. Different combination of the input (generated by the pre-processing unit) are fed to the systolic array of processing elements (PE). An example of pre-processing (PP) unit and processing element (PE) is shown in the figure 2 for two input values. Output of the pre -processing unit is the addition and subtraction of the two inputs. In the systolic array a single processing unit four inputs X_i, Y_i, U_i, V_i . Two multipliers are used in the processing element and output is U_0, V_0 . The inputs X_i and Y_i are used by the other processing elements as X_0 and Y_0 .

Pre-processor (PP)



here two multipliers α, β are used
 $U_o = U_i + \alpha * X_i$
 $V_o = V_i + \beta * Y_i$

PROCESSING ELEMENT (PE)

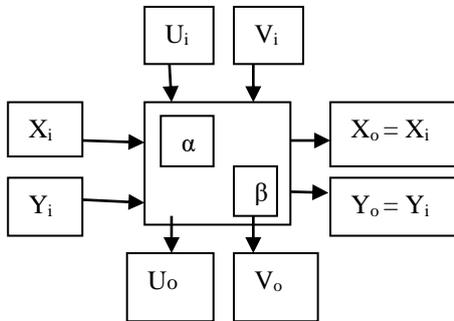


Fig2.Pre- processing unit, Processing element for two input samples

that there are N/4 pre –processing unit in stage 1 (PP1, PP2, PP3 and PP4) where all the inputs are added and subtracted to provide input to the next stage of preprocessing (PP5).

Section IV:- The idea for preprocessing unit suggested in section III is extended for N=16 .It can be seen from figure 3

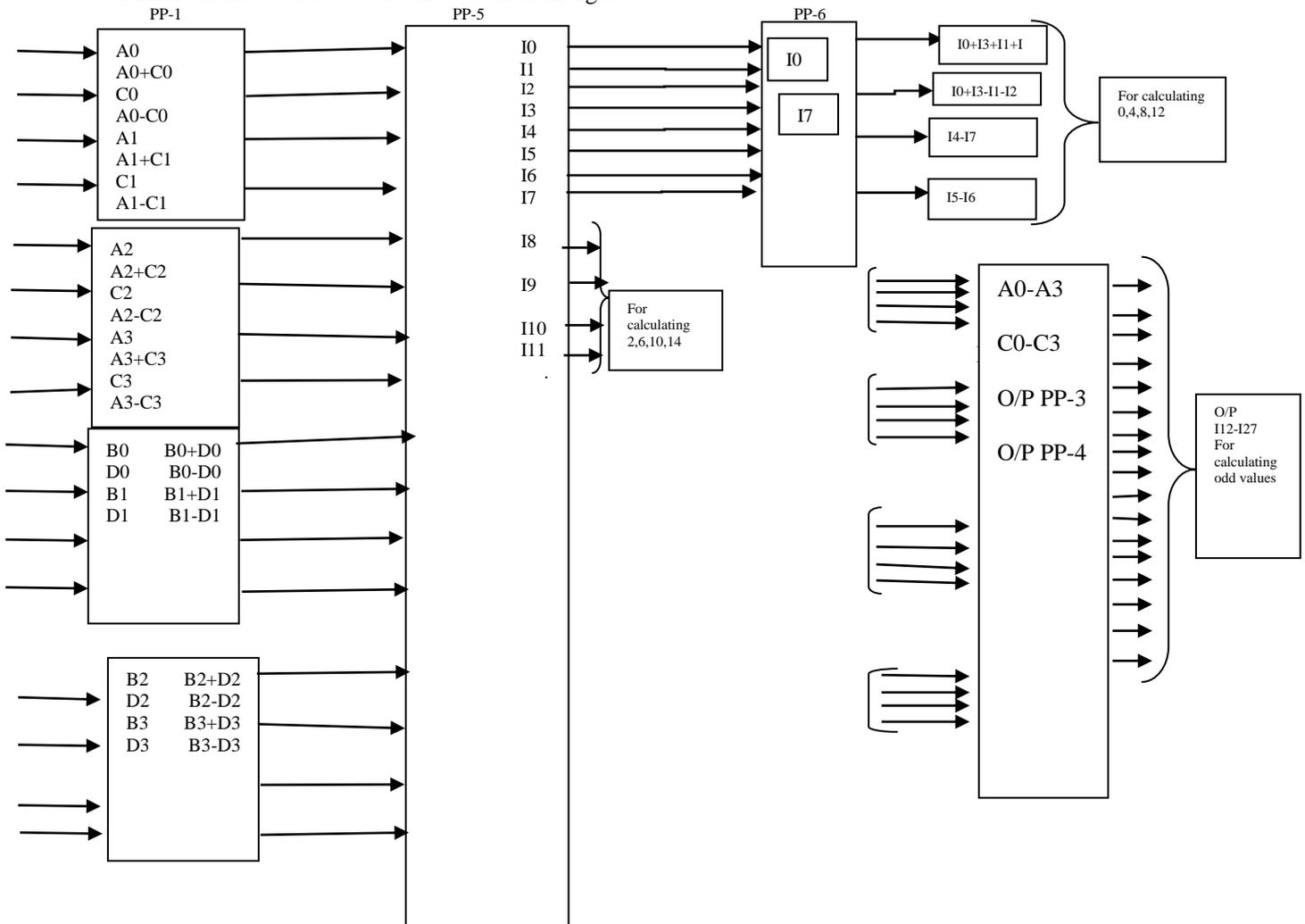


Fig3.Pre- processing unit for N=16

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O/P of PP-5 are I0 to I11 ,where

I0-I7:- $(A_n+C_n) \pm (B_n+D_n)$, $n=0$ to $N/4-1$

(I0,I1,I2,I3:- addition),(I4,I5,I6,I7:-subtraction)

I8-I11:- $(A_n-C_n) \pm (B_n+D_n)$, $n=0$ to $N/4-1$

O/P of PP-7 : I12 to I27 where

I12-I19:- $[A_n \pm (B_n-D_n)/\sqrt{2}] = [A_n \pm (1/\sqrt{2} \text{ of O/Ps from PP-3 and PP-4})]$,

I20-I27:- $[C_n \pm (B_n+D_n)/\sqrt{2}] = [C_n \pm (1/\sqrt{2} \text{ of O/Ps from PP-3 and PP-4})]$,

Eg :- I12= $A_0+(B_0-D_0)/\sqrt{2}$ I13= $A_0-(B_0-D_0)/\sqrt{2}$
I14= $A_1+(B_1-D_1)/\sqrt{2}$

I15= $A_1-(B_1-D_1)/\sqrt{2}$, I16, I17 for $n=2$, I18, I19 for $n=3$,

I20= $C_0+(B_0+D_0)/\sqrt{2}$, I21= $C_0-(B_0+D_0)/\sqrt{2}$

Systolic array of PE: The architecture is a pipelined network arrangement of Processing Elements (PEs) called cells. It is a

Figure 4 shows the systolic array unit for computation of even coefficients which are multiple of $N/4$.

specialized form of parallel computing, where cells compute the data which is coming as input and store them independently. A systolic architecture is an array composed of matrix-like rows of cells. Each cell shares the information with its neighbors immediately after processing.

For even DCT coefficients: $X[k]$ ($k=0,4,8,12$)= $X[4p+q]$ ($q=0$)

$X[0]=(I_0+I_3+I_1+I_2)*1$

$X[8]=(I_0+I_3-I_1-I_2)*(-\beta)$

$X[4]=(I_4-I_7)*\gamma+(I_5-I_6)*\alpha$

$X[12]=(I_4-I_7)*\alpha+(I_5-I_6)*\gamma$

Here $\alpha=0.3826$, $\gamma=0.9238$, $\beta=0.7071$

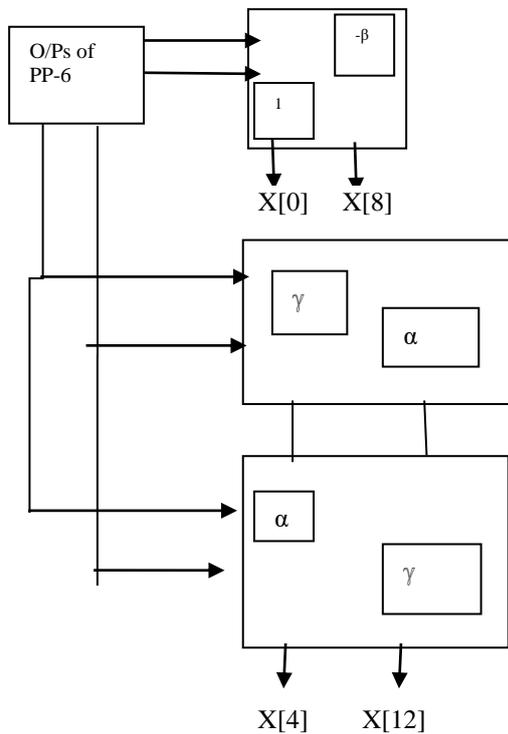


Fig 4: Processing unit for computation of multiple of $N/4$ coefficients

Remaining even coefficients $X(2,6,10,14)$, here $q=2$ are obtained as

$X[2]=(I_8*(a)+I_9*(c)+I_{10}*(-d)+I_{11}*(-b))$

$X[6]=(I_8*(c)+I_9*(-b)+I_{10}*(-a)+I_{11}*(-d))$

$X[10]=(I_8*(d)+I_9*(-a)+I_{10}*(b)+I_{11}*(c))$

$X[14]=(I_8*(b)+I_9*(-d)+I_{10}*(c)+I_{11}*(-a))$

Multiplication factors:- $a=0.9807$, $b=0.1950$, $c=0.8314$, $d=0.5555$

Systolic array of PE (one multiplier is used in these PEs)

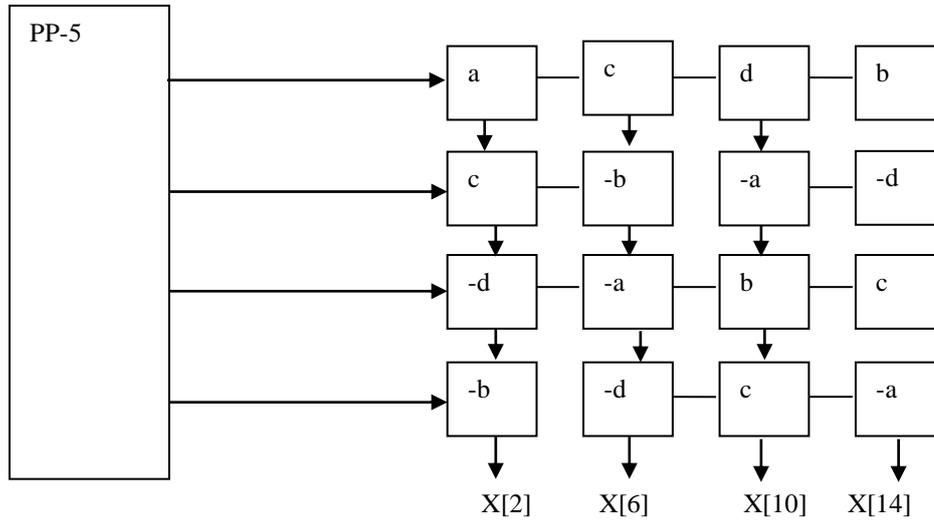


Fig 5. Processing element for computation of even coefficients of X[K]

The Processing for odd values (1, 3, 5, 7, 9, 11, 13, and 15):-

$$X[1]=m_0 \cdot I_{12} + m_2 \cdot I_{14} + m_4 \cdot I_{16} + m_6 \cdot I_{18} - m_1 \cdot I_{20} - m_3 \cdot I_{22} - m_5 \cdot I_{24} - m_7 \cdot I_{26}$$

$$X[5]=m_4 \cdot I_{13} + m_1 \cdot I_{15} - m_6 \cdot I_{17} - m_2 \cdot I_{19} - m_5 \cdot I_{21} - m_0 \cdot I_{23} - m_7 \cdot I_{25} + m_3 \cdot I_{27}$$

$$X[9]=m_7 \cdot I_{12} - m_4 \cdot I_{14} - m_3 \cdot I_{16} + m_0 \cdot I_{18} - m_6 \cdot I_{20} - m_5 \cdot I_{22} + m_2 \cdot I_{24} + m_1 \cdot I_{26}$$

$$X[13]=m_3 \cdot I_{13} - m_6 \cdot I_{15} + m_0 \cdot I_{17} - m_4 \cdot I_{19} - m_2 \cdot I_{21} + m_7 \cdot I_{23} - m_1 \cdot I_{25} - m_5 \cdot I_{27}$$

$$X[3]=m_2 \cdot I_{13} + m_7 \cdot I_{15} + m_1 \cdot I_{17} - m_5 \cdot I_{19} + m_3 \cdot I_{21} + m_6 \cdot I_{23} + m_0 \cdot I_{25} + m_4 \cdot I_{27}$$

$$X[7]=m_6 \cdot I_{12} - m_5 \cdot I_{14} - m_2 \cdot I_{16} + m_1 \cdot I_{18} + m_7 \cdot I_{20} + m_4 \cdot I_{22} - m_3 \cdot I_{24} - m_0 \cdot I_{26}$$

$$X[11]=m_5 \cdot I_{13} - m_0 \cdot I_{15} + m_7 \cdot I_{17} + m_3 \cdot I_{19} + m_4 \cdot I_{21} - m_1 \cdot I_{23} - m_6 \cdot I_{25} + m_2 \cdot I_{27}$$

$$X[15]=m_1 \cdot I_{12} - m_3 \cdot I_{14} + m_5 \cdot I_{16} - m_7 \cdot I_{18} + m_0 \cdot I_{20} - m_2 \cdot I_{22} + m_4 \cdot I_{24} - m_6 \cdot I_{26}$$

Multiplication factors: - $m_0=0.9951$, $m_1=0.0980$, $m_2=0.9569$, $m_3=0.2902$, $m_4=0.8819$, $m_5=0.4713$, $m_6=0.7730$, $m_7=0.6343$

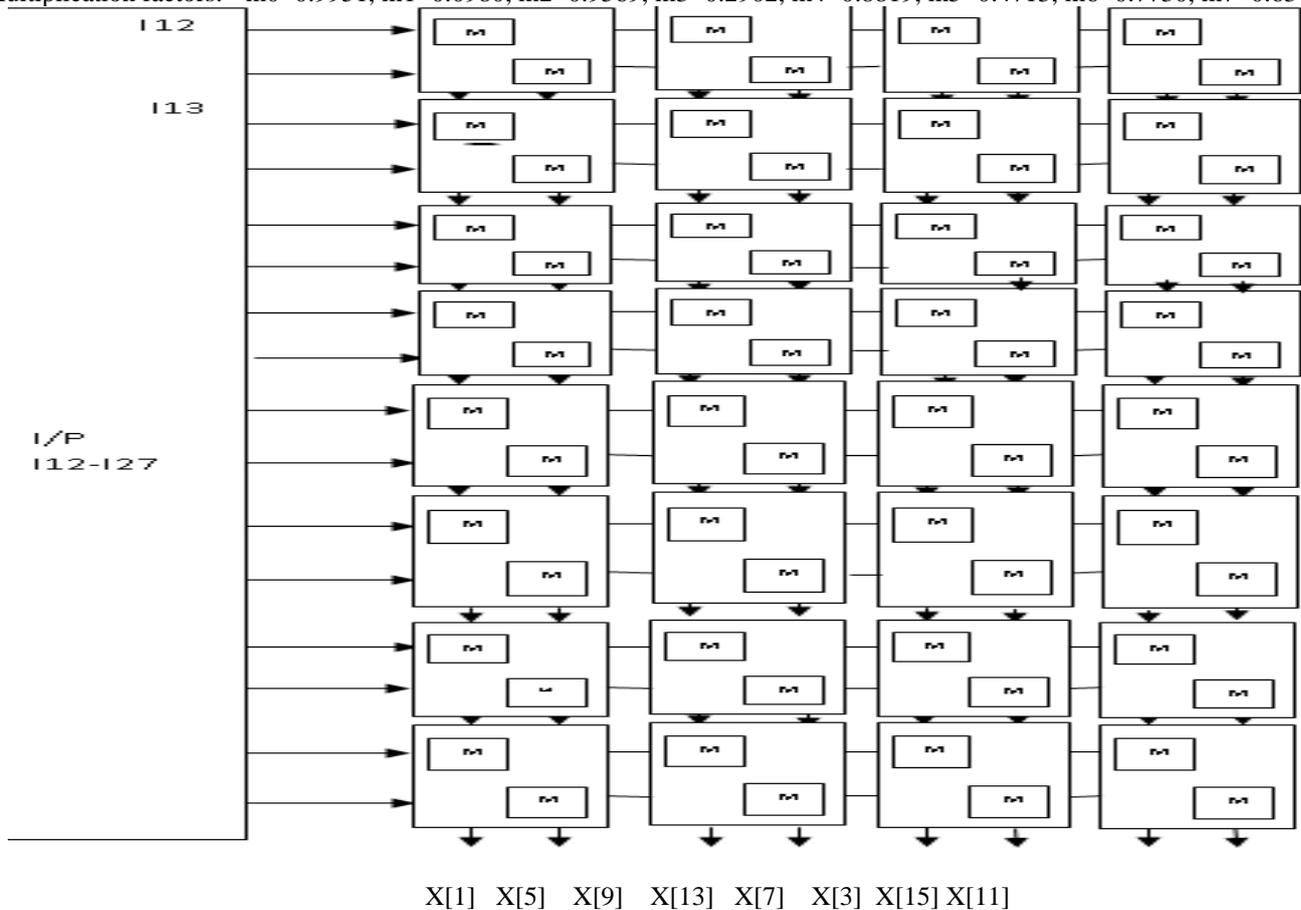


Fig-6: systolic array of PE for computing odd DCT coefficients

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Comparison: The suggested algorithm for computing DCT is a reduced computational complexity algorithm. In this algorithm the number of multiplications for $N=16$, is reduced nearly $1/3$ of the N^2 (no of multiplications required in general method). The number of PE required to calculate all the DCT coefficients of a sequence $N=16$, are 43, and the number of multipliers per PE are 2. Latency is defined as the time required to compute the DCT computations. Parallel computing of even and odd coefficients leads to a latency of $11T$ as it needs maximum time to compute the odd coefficients. For computing even coefficients the PE (with two multipliers) required are only 11, and the latency for $k=0,8$ is T , For $k=4,12$, it is $2T$, for $k=2,6,10,14$ latency is $5T$. Maximum delay is for odd values $k=1,3,5,7,9,11,13,15$, the latency for these values is $11T$. Therefore, the overall latency is consider to be $11T$. $T=(m+a)$, denotes the time required for multiplications(m) and additions(a). Table 1 shows the comparison of different systolic architecture for computation odd DCT. It can have seen from the table that the proposed architecture is efficient as it requires less number of processing elements as well as it is faster as its latency is less.

Table1. Performance comparison of systolic architecture for computing $N=16$ point DCT:

	No of real multipliers per PE	Required no. of PE	Latency	Throughput
S.B. Pan, R.H. Park[16]	2	$N^2/4=64$	$(N-1)T=15T$	T
Chang and Wang[22]	1	$N^1/2=128$	$(N-1)T=15T$	T
Proposed	2	$N^2/6=43$	$(N-5)T=11T$	T

IV. CONCLUSION:

A new algorithm for computing DCT and its VLSI implementation using parallel and pipelined network of processing elements is proposed in the paper. The array of processing elements works in systolic fashion which makes the computation fast.

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